Achieving Fast Switching with a High-Voltage Isolated Driver IC

Industrial systems have struggled in recent years with the need of high efficiency, combined with the need of more compact systems. This also influences the choice of the gate drivers associated to high speed switches, like Silicon Carbide (SiC) MOSFETs. STMicroelectronics has recently launched the STGAP2, a family of isolated gate drivers focused on the industrial power market. Its portfolio includes different isolation voltages and functionalities.

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Isolated gate drivers are widely used in industrial power applications. In high switching frequency applications, the high commutation slopes of the power switches require immunity against high levels of common mode noise. In addition, isolation capability and very short signal propagation delay are also typically required.

The STGAP2S product family [1] [2], from STMicroelectronics, offers isolation through a coreless transformer, with propagation delay of 80 ns. In addition, it has an immunity against common mode above 100 V/ns. The isolation between primary and secondary sides is tested up to 1.7 kV. All this in a very compact package: the single channel driver is available in SO-8N package, while the STGAP2D - dual channel – is available in SO-16N package. The is offered in two variants: one STGAP2SCM, with active Miller clamp (AMC), and STGAP2SM, with separated on- and off- outputs. The comparative table of all devices from STGAP2 technology is shown below in Table 1.

The devices from the STGAP2S family have 4 mm creepage distance between primary and secondary sides. For some applications, there might be the need of higher creepage distances, depending on pollution degree and maximum humidity levels.

In these cases, STGAP2HS [3] family is indicated. It is offered in an SO-8W package, with 8 mm creepage, allowing the gate drivers from this family to be tested with 6 kV.

In addition, this family also includes dedicated part numbers for operation with SiC MOSFETs, namely STGAP2SiCS [4]. They have increased values of under voltage lockout (UVLO) to couple with the higher operation gate voltage of those switches.

Common Mode Transient Immunity

Fast power switches applied in industrial applications can present slew rates above 50 V/ns. This is a desired feature to reduce dynamic losses and achieve higher system efficiency. Fast transients can generate disturbances in the associated gate driver though. In high side drivers, the fast transient in the secondary side can generate oscillations in the primary ground. In the worst case, this may cause glitches in the input signal, and consequently, the unwanted switch on of the SiC MOSFET. Figure 1 shows experimental results with the STGAP2S device during very fast transients, both positive and negative, with a DC voltage V_{dc} =1500 V. The waveforms demonstrate that, even for slew rates in the order of 120 to 130 V/ns, the gate driver was able to properly operate and maintain the desired output state.

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Coupling Through Miller Capacitance

High slew-rate transients in the Drain to Source voltage can also cause oscillations in the gate through the Miller capacitance (MC) i.e. the parasitic capacitance between Drain and Gate of a MOSFET. To

2. Part number	# Channels	Package	Isolation ¹	UVLO ²	Output Conf.
STGAP2SM	Single	SO-8N	1.7 kV	9.1 V	Separated on/off
STGAP2SCM	Single				Miller Clamp
STGAP2D	Dual	SO-16N			Single Output
STGAP2HSM	Single	SO-8W	6 kV	9.1 V	Separated on/off
STGAP2HSCM					Miller Clamp
STGAP2SiCS				15.5 V	Separated on/off
STGAP2SiCSC					Miller Clamp

Table 1: Product portfolio of STGAP2 Technology.

¹ more details on the isolation voltage in a dedicated section.

referred as the typical value of the turn-on threshold, as stated in the datasheet.



Figure 1: Measured positive and negative transients in the STGAP2S, V_{DC}=1500V.

minimize it, one of the variants of the STGAP2S, the STGAP25CM, offers an active Miller clamp (AMC), see Figure 2.

The AMC is a recommended functionality for use with SiC MOSFETs in applications where fast hard-switching transients are expected. Figure 3 depicts idealized waveforms of a half-bridge inverter using the SCTW35N120G2V, a 650V rated SiC MOSFET from STMicroelectronics [6]. As the output current is assumed positive, the low side





Figure 2: Available options of the STGAP2S.





Figure 3: Half-Bridge inverter with SIC MOSFETs, and idealized waveforms of V_{DS} and V_{GS}.







Table 2: Voltage characteristics of the STGAP2H devices in SO-8W package.

switch (S2) is turned-on during the free-wheeling time. After a short dead-time after the turn-off of S2, the high side switch S1 is turnedon. When this happens, the V_{DS} of S1 goes to zero, causing the V_{DS} of S2 to increase. Both transients have same slew-rate. The gate voltage V_{GS} of S2, which was in OFF voltage, is now pulled up by the coupling through Miller capacitance. In the same way, V_{GS} is pulled down when S1 turns off. Symmetrically, the same kind of oscillations are expected in S1 when output current is negative.

Both positive and negative oscillations in V_{GS} during OFF state can have a negative impact on the device and the system. Positive oscillations can cause parasitic turn-on of the MOSFET, and consequent shot-thru of the entire half-bridge. On the other hand, negative oscillations may bring V_{GS} outside of the safe operation area (SOA), and trigger degradation mechanisms in the device.

As can be seen by the waveforms in Figure 4, the ACM is able to reduce both positive and negative spikes of V_{GS} during off state, down to safe values. As it can be shown, when driven by the STGAP2SM, the V_{GS} of a SiC MOSFET can achieve positive values, with the consequent risk of parasitic-turn-on. With the STGAP2SCM, the same configuration presents much lower spikes, which never reach positive values.

In the same way, the negative oscillations in the right side of Figure 4 are also reduced. The configuration with the STGAP2SM presents negative peaks below the absolute minimum gate voltage, which lays at $V_{GS,min}$ =-10 V for the SCTW35N65G2V. With the STGAP2SCM, the negative spikes never reach $V_{GS,min}$.

Isolation voltage

The gate drivers inside the SO-8W package – refer to Table 1 – have an isolation of 6 kV. This voltage is related to the test voltage during production, in accordance to UL1577. More details on the voltage characteristics of the STGAP2H family in SO-8W package are shown in Table 2.

Undervoltage Lock-out (UVLO)

The undervoltage lock-out (UVLO) is a protection feature, present in all STGAP2 devices. It prevents the power switch from being driven with a voltage below its requirements. The UVLO protection is activated when the supply voltage of the secondary side, i.e. the voltage between pins VH and GNDISO, drops below a certain value.

IGBTs and super-junction MOSFETs operate with gate voltages between +12 V and +15 V. Below this range, the on-resistance of the MOSFET – or the saturation voltage of the IGBTs – starts to increase, and with it the conduction losses of the switch. The switch could also start to operate in linear mode, leading to thermal runway and to device failure. The same can also occur in SiC MOSFETs. However, due to the required higher voltage on the gate – from +18V to +20V, - the value at which the UVLO is activated needs to be increased accordingly.

Figure 5 presents the output characteristics of the STW90N65G2V [5]. The V-I curves on the graph are obtained for different values of gate-to-source voltages V_{GS} , at room temperature. The values below the graph are the calculated power dissipation due to current conduction, at a drain current of 40 A. For V_{GS} = 18 V, the nominal value for this device, the resulting dissipation is around 28 W.



Figure 5 – Output characteristics of SCTW90N65G2V SiC MOSFET, and calculated conduction losses for different gate-tosource voltages.

	Min	Тур.	Мах				
$V_{HL(ON)}[V]$	14.6	15.5	16.4				
$V_{HL(OFF)}[V]$	13.9	14.8	15.7				

If the power supply on the gate driver drops, this will affect V_{GS} as well. On the graph we have the example for V_{GS} = 12 V, that would already increase the conduction losses almost double. Going even further down will cause the MOSFET to operate in linear mode. The extremely high associated losses can cause the MOSFET to fail due to overheating.

To avoid that, the UVLO of the STGAP2SiCS [5] devices has been increased to 15.5 V. This guarantees proper protection even if bipolar drive is applied to the MOSFET. As an example, if a negative off voltage of -3V is applied to the MOSFET, this will bring the effective activation voltage down to +12.5V, still safe enough for SiC MOSFETs.

Summary

The devices in the STGAP2 technology family enable operation at very high switching speeds, and in turn, enabling high power density of industrial systems. Among their very interesting features are CMTI above 100 V/ns, typical propagation of 80 ns, and active Miller clamp functionality. The newest STGAP2HS is offered in an SO-8W package, which allows a higher isolation voltage of 6 kV. Specifically designed for combination with SiC MOSFETs, the STGAP2SiCS has an increased UVLO value, to guarantee its proper protection function.

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